

### **Current IC, Built-in Magnetic Converter for Sensing Horizontal Magnetic Fields**

# **MagnTek**

#### Now Part of **NOVOSENSE**

## **1 Product Description**

The MagnTek® MT9519 product series is a current IC that utilizes a built-in magnetic collector to convert a vertical magnetic field into horizontal magnetic dection (HMD). The traditional Horizontal Hall technology is only sensitive to the magnetic flux density applied perpendicular to the IC surface. The sensor IC using HMD technology is only sensitive to the magnetic flux density applied parallel to the surface of the IC. This is achieved by placing two pieces of magnetic conductive metal on the chip. This chip is a highly integrated Hall sensor IC that provides an output signal proportional to the magnetic flux density applied horizontally, making it suitable for current measurement. Due to its small size and use of SOP-8 package, it is very suitable as an open-loop current sensor for PCB or busbar installation. It is suitable for various current ranges from a few amperes to 1000 amperes or higher. MT9519 is user programmable, including current direction, magnetic field sensitivity gain, zero magnetic field signal output, and temperature compensation. It is very suitable for the application of on-board inverters.

The MT9519 series provides customers with SOP-8 package that meets RoHS requirements.

### **2 Features**

- End-of-line programmable
- **•** Typical Accuracy: **---** ±1.0%(25℃)
- **■** High Linearity:
	- **---** ±0.5%(25℃)
- High Bandwidth: **---** 250kHz
- Wide Operating Temperature: **---** -40℃~150℃
- Fast Output Response Time: **---** 2.2 μs (typ.)
- Package Option: ---SOP-8
- High stability over operation temperature range: ---±1.5%( 25℃~125℃)
- ---±1.5%(-40℃~25℃)
- Ratiometric Output from Supply Voltage
- Low-Noise Analog Signal Path
- RoHS Compliant: (EU)2015/863

### **3 Applications**

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- Vehicle mounted drive motor inverter
- PV string inverters
- Battery management system
- Switching power supplies
- Overcurrent protection



### **4 Product Overview of MT9519**



*Not to scale* 

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1 Originally Version

### **5 Functional Block Diagram**



# **6 Pin Configuration and Functions**





Figure.2 Pin Configuration and Functions (SOP-8)



### **7 Naming Specification**



### ❶ Series Name

### ❷ Package Type



### <sup>3</sup> Sensitivity Range



### **8 Selection Guide**



### **9 Typical Application Circuit**

The typical application circuits of MT9511series products include a bypass capacitor and a filter capacitor as an additional external components. **CBYPASS capacitor between VCC and GND is necessary.** Magnetic field applied horizontally to chip surface, the analog signal output is measured directly from the VOUT pin



### **10 Transfer Characteristics**





Figure.4 Output Voltage vs. Magnetic Range

The induction direction of the MT9519 series product is shown in the Figure.5. The current direction flowing through the copper bar is shown as IP. The magnetic field direction line generated by the magnetic focusing plate is shown as B.





Figure.5 Current Direction & Output Polarity

# **11 Electrical Magnetic Characteristics**

### **11.1 Absolute Maximum Ratings**

Absolute maximum ratings are limited values to be applied individually, and beyond which the serviceability of the circuit may be impaired. Functional operability is not necessarily implied. Exposure to absolute maximum rating conditions for an extended period of time may affect device reliability.



### **10.2 ESD Ratings**



 $\mathcal{E}$ 

## **11.3 Electrical Specifications**

TA =-40~125 ℃, VCC=5V, CBYPASS=0.1uF (unless otherwise specified)



Continued on the next page…

# **Electrical Specifications(continued)**





Accuracy Specification



Continued on the next page…

## **Electrical Specifications(continued)**

TA=-40~125 ℃, VCC=5V, CBYPASS=0.1uF (unless otherwise specified)





UNES

### **12 Characteristic Definitions**

#### **Power On Time---TPO**

When the supply is ramped to its operating voltage, the device requires a finite time to power its internal components before responding to an input magnetic field.

The Power-On Time (TPO) is defined as the time taken between the supply reaching the minimum operating voltage VCCmin (t1), and the output voltage to settling to within  $\pm 10\%$  of its steady state value under an applied magnetic field (t2) (See Figure 6).



Figure.6 **Power On Time Definition**

#### **Propagation Delay---TPD**

The time interval between a) when the primary current signal reaches 20% of its final value, and b) when the output reaches 20% of its final value (see Figure.7).



#### **Rise Time---TR**

Rise Time is the time interval between the sensor VOUT reaching 10% of its full scale value (t1), and it reaching 90% of its full scale value (t2). (see Figure.8). Both TR and TRESP can be negatively affected by any eddy current losses created if a conductive ground plane is used.



#### **Response Time---TRESP**

The time interval between a) when the primary current signal reaches 80% of its final value, and b) when the sensor reaches 80% of its output corresponding to the applied current. (see Figure.9). Both TR and TRESP can be negatively affected by any eddy current losses created if a conductive ground plane is used.



#### **Delay to Clamp---TCLP**

A large magnetic input step may cause the clamp to overshoot its steady state value. The Delay to Clamp (TCLP) is defined as the time it takes for the output voltage to settle within  $\pm 1\%$  of its steady state value, after initially passing through its steady state voltage (see Figure.10)。



#### **Quiescent Voltage Output---VOQ**

In the quiescent state (no significant magnetic field:  $B = OGS$ ), the output (VOQ), has a constant ratio to the supply voltage (VCC), throughout the entire operating ranges of VCC and ambient temperature (TA), VOQ=0.5\*VCC。

#### **Quiescent Voltage Output Drift Through Temperature Range---∆VOQ\_TC**

Due to internal component tolerances and thermal considerations, the Quiescent Voltage Output (VOQ), may drift from its nominal value through the operating ambient temperature (TA). The Quiescent Voltage Output Drift Through Temperature Range, *∆* VOQ\_TC, is defined as:

*∆VOQ\_TC=VOQ(TA)−VOQ\_EXPECT(TA)*

VOQ\_TC should be calculated using the actual measured values of VOQ(TA) and VOQ\_EXPECT(TA) rather than programming target values

#### **Sensitivity---SNST**

The magnetic field horizontal to the packaging marking surface is linearly related to the output voltage. The larger the magnetic field, the greater the change in output voltage, and vice versa, the smaller the change in output voltage. This ratio is specified as the magnetic sensitivity SNST (mv / Gs) of the chip, is defined as:

> *SNST= VOUT(BPOS)<sup>-V</sup>OUT(BNEG) BPOS−BNEG*

where BPOS and BNEG are two magnetic fields with opposite polarities.

#### **Sensitivity Drift Through Temperature Range---∆SNST\_TC**

The temperature coefficient effect of sensitivity can cause magnetic sensitivity to deviate from its expected value in the operating ambient temperature range (TA). The Sensitivity Drift Through Temperature Range, *∆* SNST\_TC, is defined as:

∆SNST\_TC*= SNST(TA)−SNST\_EXPECT(TA) SNST\_EXPECT(TA)* <sup>∗</sup>*100%*

#### **Sensitivity Drift Due to Package Hysteresis---ΔSNST\_PKG**

The stress effect during packaging can cause magnetic sensitivity to deviate from its expected value. The Sensitivity Drift Through Temperature Range, ΔSNST\_PKG, is defined as:

$$
\Delta SNST\_PKG \texttt{=}\frac{SNST\_25\text{°}C\_2\texttt{-}SNST\_25\text{°}C\_1}{SNST\_25\text{°}C\_1} * 100\%
$$

where SNST\_25 ℃ \_1 is programmed value of sensitivity at TA=25℃, and SNST\_25℃\_2 is the value of sensitivity at TA=25℃, after temperature cycling from TA to 150℃/168 hours and back to 25℃

#### **Nonlinearity Sensitivity Error---ELIN**

Ideally input magnetic field vs sensor output function is a straight line. The non-linearity is an indication of the worst deviation from this straight line. The ELIN in % is defined as:

$$
ELIN = \left(\frac{SNST\_B1}{SNST\_B2} - 1\right) * 100\%
$$

Where:

$$
SNST\_B1=\left(\frac{VOUT\_BPOS1-VOUT\_BNEG1}{BPOS1-BNEG1}\right)
$$

$$
SNST\_B2 = \left(\frac{VOUT\_BPOS2-VOUT\_BNEG2}{BPOS2-BNEG2}\right)
$$

and BPOSx and BNEGx are positive and negative magnetic fields, with respect to the quiescent voltage output such that |BPOS2| = |BNEG2| = Bmax, and  $|BPOS2| = 2 * |BPOS1|$  and  $|BNEG2| = 2$ \* |BNEG1|.

#### **Symmetry Sensitivity Error---ESYM**

The magnetic sensitivity of MT9519 device is constant for any applied magnetic fields of equal magnitude and opposite polarities. Symmetry Error (ESYM) is measured and defined as:

*ESYM= SNST\_BPOSx SNST\_BNEGx −1* <sup>∗</sup>*100%*

Where:

*SNST\_BPOSx= VOUT\_Bx−VOQ Bx*

*SNST\_BNEGx= VOQ−VOUT\_Bx Bx*

BPOSx and BNGx are positive and negative magnetic fields such that  $|BPOSx| = |BNEGx|$ .

#### **Ratiometry Error---ERAT**

The MT9519 device features ratiometric output. This means that the Quiescent Voltage Output (VOQ), magnetic sensitivity (SNST) and Output Voltage Clamp (VCLP\_HI, VCLP\_LO), are proportional to the Supply Voltage (VCC). In other words, when the VCC increases or decreases by a certain percentage, each characteristic also increases or decreases by the same percentage. Error is the difference between the measured change in the VCC relative to 5V, and the measured change in each Characteristic

#### **Ratiometry Quiescent Voltage Output Error---ERAT\_VOQ**

ERAT VOQ, for a given supply voltage, is defined as:

$$
ERAT\_VOQ = \Big(\frac{VOQ(VCC)/VCC)}{VOQ(5V)/5V} - 1\Big) *100\%
$$

#### **Ratiometry Sensitivity Error--ERAT\_SNST**

ERAT SNST, for a given supply voltage, is defined as:

$$
ERAT\_SNST = \left(\frac{SNST\_B1(VCC)/VCC)}{SNST\_B1(5V)/5V} - 1\right) *100\%
$$

**Ratiometry Clamp Error---ERAT\_CLP**

ERAT CLP, for a given supply voltage, is defined as:

$$
ERAT\_CLP = \left(\frac{VCLP(VCC)/VCC)}{VCLP(5V)/5V} - 1\right) *100\%
$$

Where VCLP is either VCLP HI or VCLP LO.

#### **Over Current Limit---ISCLP & ISCLN**

The MT9519 has over-current protection function. When IOUT≥ISCLP or ISCLN, the output driver will be closed and the output will be turned into high resistance state.

### **Power-On Reset---POR, Undervoltage Lockout---UVL**

The descriptions in this section assume temperature = 25°C, no output load (RL, CL) , and no significant magnetic field is present.

Power-Up. At power-up, as VCC ramps up, the output is in the following power supply voltage state. When VCC exceeds VPORH, the chip will enters the handshake protocol state. When VCC exceeds VUVLOH, the output will go to 1/2\*VCC or 2.5V, at this time, the chip is in normal working state.

Power-Down. If VCC drops below VUVLOL, the output will be in a high-impedance state. If VCC drops below VPORL, the output is in the following power supply voltage state (See Figure. 11).



Figure.11 **POR and UVL Definition**

# **13 Package Material Information (For Reference Only – Not for Tooling Use)**

# **13.1 SOP-8 Package Information**





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